(Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Numerical filter code analysis Frama-C Days 2024

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²EDF Lab Chatou - PRISME - P11

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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- (2) 1st order filter
- ilter with real eigenvalues
- ④ filter with complex eigenvalues





Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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- 2 1st order filter
- ③ filter with real eigenvalues
- ④ filter with complex eigenvalues
- **5** Conclusion



Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Numei	rical filter			

- A (discrete-time) signal x : Z → R maps discrete times to real values.
- A filter ℋ: (Z → R) → (Z → R) maps an input signal u into an output signal y (or vector of signals).

$$\stackrel{u(k)}{\longrightarrow} \stackrel{y(k)}{\mathcal{H}} \xrightarrow{}$$

• We restrict here to Linear Time Invariant (LTI) filters of finite order *n*. Its canonical form is a constant-coefficient difference equation.

LTI filter

$$y(k) = \sum_{i=0}^{n} b_i . u(k-i) - \sum_{i=1}^{n} a_i . y(k-i)$$

where $\{a_i\}_{1 \le i \le n}$ and $\{b_i\}_{0 \le i \le n}$ are constant real coefficients.

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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EDF c	ontext			

- Qualification of critical control systems for nuclear power plant
- We must verify the absence of RTE (*RunTime Error*)
- In particular, the absence of numerical overflows
- A control program classically starts with the filtering of sensor signals

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- So the proven filter bounds should be as accurate as possible!
- Toy example :

```
y = filter(input);
//@ assert y <= 49.9;
output = 1.0 / (50.0 - y);</pre>
```

Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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State o	of the art			

- More than two decades of effort by the abstract interpretation community, that produced a number of tuned **relational numerical abstract domains**, implemented in Astree or Polyspace:
 - Ellipsoids [2, 3, 10]
 - Zonotopes [1]
 - Set of invariants (Boxes, zonotopes, polyhedra...) [7, 9]
- None of them are currently mature in Frama-C
- Current approch:
 - Pre-compute the invariant once and for all outside Frama-C
 - Import the invariant as an ACSL annotation
 - Oheck the invariant within Frama-C (Eva or Wp)
- By the way: recent breakthrough in the arithmetic community [5]

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Bound	ling filter o	utput		

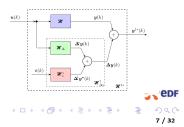
Worst-Case Peak-Gain (WCPG) Theorem

o

 $\|\mathscr{H}(u)\|_{\infty} \leq \|\mathbb{h}\|_{1}.\|u\|_{\infty}$

where $\mathbb{h} = \mathscr{H}(\delta)$ is the **impulse response** of the filter (δ is the impulse signal: $\delta(0) = 1$ and $\forall k \neq 0, \delta(k) = 0$).

- ||h||₁ is called the Worst-Case Peak-Gain (WCPG) and can be correctly computed [12].
- This bound is optimal and reachable [4].
- To take into account encoding error, we must introduce two additional error filters: ℋ_Δ (coefficient quantization) and ℋ_ε^{*} (roundoff error), where ε is the error signal depending on the implementation details (algorithm, machine arithmetic...) [5].



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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Warm	up			

Example of 1st order filter

$$\left\{ egin{array}{ll} orall k & u(k) \in [-10,10] \ orall k < 0 & y(k) = 0 \ orall k \geq 0 & y(k) = 0.9y(k-1) + 0.5u(k) \end{array}
ight.$$

$$\begin{split} \mathbb{h}(k) &= 0.5 * 0.9^k \\ \implies & \|\mathbb{h}\|_1 = 0.5 \sum_{k=0}^{\infty} 0.9^k = 0.5 \frac{1}{1 - 0.9} = 5 \\ \implies & \|y\|_{\infty} \le 5 \|u\|_{\infty} = 50 \end{split}$$

 $\bullet \ 1^{st}$ order filters can be abstracted in interval domain:

$$0.9 * [-50, 50] + 0.5 * [-10, 10] = [-50, 50]$$

• This is also correct in floating point arithmetic !

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Demo	1			

```
double out1;
const double ki[] = \{0.5\};
const double ko[] = \{ 1.0, -0.9 \};
int main() {
  double in=0, out=0;
  //@ widen_hints out, -50.0, 50.0;
  while (Frama_C_interval(0, 1)) {
    out1 = out:
    in = Frama_C_double_interval(-10.0, +10.0);
    out = -ko[1]*out1 + ki[0]*in;
  }
The command
$ frama-c -eva filter1.c
produces the following output:
[eva] ===== VALUES COMPUTED ======
[eva:final-states] Values at end of function main:
  y \in [-50. .. 50.]
```

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Increa	sing the co	mplexity		

Example of 2nd order filter with real eigenvalues

$$\begin{cases} \forall k & u(k) \in [-10, 10] \\ \forall k < 0 & y(k) = 0 \\ \forall k \ge 0 & y(k) = y(k-1) - 0.2 * y(k-2) + 0.1 * u(k) \end{cases}$$

The characteristic polynom is $p(\lambda) = \lambda^2 + \lambda - 0.2$. We can check that its discriminant is $\Delta = 0.2$ and so its eigenvalues are real:

- $\lambda_1 pprox 0.27639320225 \ \lambda_2 pprox 0.72360679775$
- WCPG theorem application¹: $\|y\|_{\infty} \leq 0.5 * \|u\|_{\infty} = 5$
- Interval domain does not help anymore !

$$[-5,5] - 0.2 \ast [-5,5] + 0.1 \ast [-10,10] = [-8.5,8.5] \not\subseteq [-5,5]$$

¹WCPG can be safely computed using https://github.com/fixif/WCPG <=>

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Linear decomposition theorem

A filter of order n having only real eigenvalues can be decomposed into a linear combination of n filters of order 1.

$$y = \frac{\lambda_1 \cdot e_1 + \lambda_2 \cdot e_2}{\lambda_1 - \lambda_2}$$

with

$$\begin{cases} e_1(k) = \lambda_1 \cdot e_1(k-1) + 0.1 * u(k) \\ e_2(k) = \lambda_2 \cdot e_2(k-1) - 0.1 * u(k) \end{cases}$$

- We can implement this implementation as ghost code and prove it with Wp and external solvers.
- Eva finish the proof to bound e_1 , e_2 and then y.
- Limitation: the Wp proof rely on the real model !
- Perspective: use an external mechanized filter theory (like in [4]) to bound the floating-point error and import the proof in Wp .

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Demo	2			

```
The command

$ frama-c -wp -wp-prover altergo,z3 -wp-model real

filter2-fc.c -then -eva filter2-fc.c

produces the following output:
```

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Example of 2nd order filter with complex eigenvalues

$$\left\{ egin{array}{ll} orall k & u(k) \in [-10,10] \ orall k < 0 & y(k) = 0 \ orall k \geq 0 & y(k) = 1.5 * y(k-1) - 0.75 * y(k-2) + 0.5 * u(k) \end{array}
ight.$$

- We can check that $\Delta = -0.75 < 0$
- WCPG theorem application: $\|y\|_{\infty} \leq 4.91892 \|u\|_{\infty} = 49.1892$
- For this kind of filter, we have to generate relational invariants:

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- ellipsoid
- ${\scriptstyle \bullet}$ set of boxes
- zonotope
- fractal zonotope

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Ellipsoï	ď			

• Idea: Find an invariant of the form [3, 10]:

$$y(k)^{2} + a * y(k-1)^{2} + b * y(k) * y(k-1) + c * y(k) + d * y(k-1) \le e$$

• Invariant found for our toy example:

$$y(k)^{2} + 0.75 * y(k-1)^{2} - 1.5 * y(k) * y(k-1) \le 1239.81$$

- Advantages: short, readable and verifiable by hand invariant
- Drawbacks:
 - Invariant verification with Wp is hard [8]
 - rely on an external polynomial solver
 - don't work if the norm of the complex eigenvalues are too close to 1
 - worst accuracy $\longrightarrow |y(k)| \le 81.3162 \text{ (recall: } \|y\|_{\infty} = 49.1892 \text{)}$

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Set of	boxes			

- Idea: state partitioning abstraction to compute an invariant of kind disjunction of intervals [9]
 - Divide and conquer (split the input and state intervals in many chunks)
 - The trick comes from constraint programming research field
- Advantages:
 - fully automatic: no need for a priori knowledge
 - \bullet verifiable with Eva within interval domain \longrightarrow natively valid in floating-point arithmetic

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• Drawbacks:

- $\bullet\,$ bad performance \longrightarrow generation : 4s , verification : 62s
- better but still bad accuracy $\longrightarrow |y(k)| \le 65$ (recall:
 - $||y||_{\infty} = 49.1892$

Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Zonoto	ope			

• Idea: Find an invariant in an affine form [1]:

$$\bigwedge_{k=0}^{n} y(k) \in \alpha_{k,0} + \sum_{i=1}^{N} \alpha_{k,i} * \varepsilon_{i}$$

where $\forall i, \varepsilon_i \in [-1, 1]$ are shared **noise symbols**

• at each iteration, instead of applying classic widening, we introduce a symbolic perturbation over the affine equations, until a post-fixpoint is reached.

• Advantages:

- good accuracy $\longrightarrow |y(k)| \leq 50.32$ (recall: $\|y\|_{\infty} = 49.1892$)
- relatively basic reasoning to prove the invariant (linear simplification + interval arithmetic)

Drawbacks:

- rely on external complex algorithms and heuristics (Fourier-Motzkin, Simplex)
- ullet don't work if the norm of the complex eigenvalues are too close to 155 COF

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues 00000000	Conclusion 0000000000
Fractal	zonotope			

• Idea: expliciting the raw relation between y and u, through the impulse response h:

$$y(k) = (\mathbb{h} * u)(k) = \sum_{i=0}^{k} \mathbb{h}(i).u(k-i)$$

- \bullet Remark: the impulse response \mathbbm{h} is an inductive sequence
- the associated invariant is a fractal zonotope (a kind of "zonotope with infinite vertices").

• Advantages:

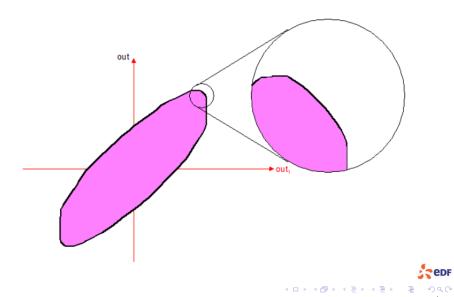
- find the optimal bound
- work in the general case
- the invariant is automatically computed, without hints

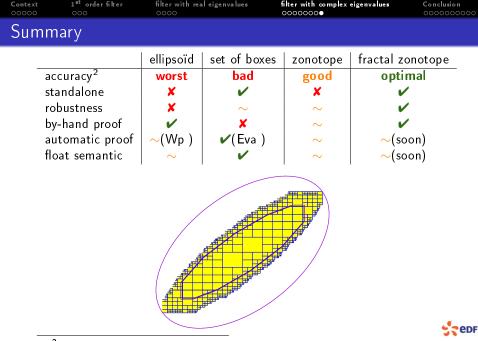
Drawbacks:

 cannot be automatically verified within Frama-C yet (soon within the Numerors plugin [6])

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²Every method is accurate enough to prove that the filter itself does not overflow

Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Take a	way			

- Control program starts by filter \longrightarrow need for accurate analysis!
- Arithmeticians are tackling the challenge to design correct-by-construction filters (WCPG theorem + sound error analysis) [5, 4, 12] → reached for fixed-point implementations
- Eva can prove 1st order filter optimal bounds
- For higher order filter, arbitrary accurate bounds can be automatically obtained within Frama-C using **subdivision techniques**...but it's costly!
- work-in-progress: implementing the fractal zonotope in Frama-C to infer automatically and efficiently optimal bounds.
- more details on the Frama-C book's chapter 12 and the companion document [11]

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Keep	going			

- Merge effort from arithmetic and abstract interpretation communities.
- Filter detection: manage the diversity of implementations
- Next challenge: mutable filter

```
int main() {
  double y, y1, y2, u=0;
while (Frama_C_interval(0, 1)) {
    u = Frama_C_double_interval(IN_MIN, IN_MAX);
    fast_convergence = Frama_C_interval(0, 1);
    y^2 = y^1;
   y1 = y;
   if (fast_convergence) {
    y = a11 * y1 + b11 * u;
    }
    else {
      y = a21 * y1 + a22 * y2 + b21 * u ;
    }
```

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Annex





In our toy example, we have:

$$\left\{egin{array}{ll} \mathbb{h}(0) = 0.5 \ \mathbb{h}(1) = 0.75 \ orall k & \mathbb{h}(k) = 1.5*\mathbb{h}(k-1) - 0.75*\mathbb{h}(k-2) \end{array}
ight.$$

We can check that \mathbb{h} is pseudo-periodic, so the optimal invariant is a classic 12-faces zonotope, described by the following formula (see [11] for details):

$$\begin{cases} y(k) \in \frac{\|u\|_{\infty}}{1-\frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3}-k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) \in \frac{\|u\|_{\infty}}{1-\frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2}-k\frac{\pi}{6}\right) \varepsilon_{k} \end{cases}$$

where $\forall k \in [0, 11], \varepsilon_k \in [-1, 1].$

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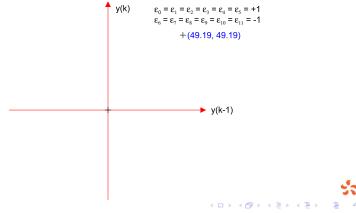
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Graphical construction of the zonotope

$$\begin{cases} y(k) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{cases}$$



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Graphical construction of the zonotope

$$\begin{cases} y(k) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{cases}$$

y(k) $\varepsilon_6 = \varepsilon_1 = \varepsilon_2 = \varepsilon_3 = \varepsilon_4 = \varepsilon_5 = +1$ $\varepsilon_0 = \varepsilon_7 = \varepsilon_8 = \varepsilon_9 = \varepsilon_{10} = \varepsilon_{11} = -1$ ▶ y(k-1) <ロ> <同> <同> <同> < 同> < 同> < □> <

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Graphical construction of the zonotope

$$\begin{cases} y(k) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{cases}$$

 ϵ_1 : +1 \rightarrow -1 $\epsilon_7: -1 \rightarrow +1$ x += -17.30 (49.19, 31.89)y += -25.95 (31.89, 5.95) → y(k-1) イロト イヨト イヨト イヨト



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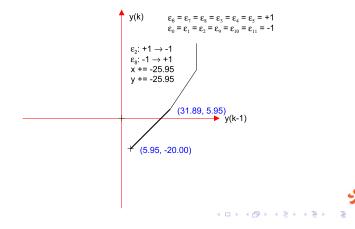
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Graphical construction of the zonotope

$$\begin{cases} y(k) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{cases}$$



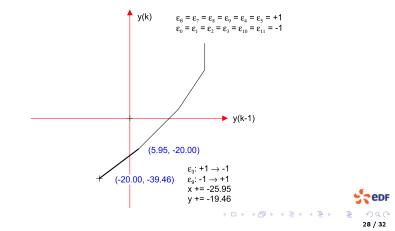
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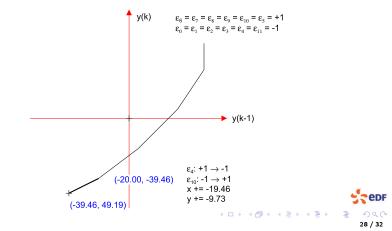
$$\begin{array}{rcl} y(k) & \in & \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) & \in & \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{array}$$



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$$\begin{array}{rcl} y(k) & \in & \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) & \in & \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{array}$$

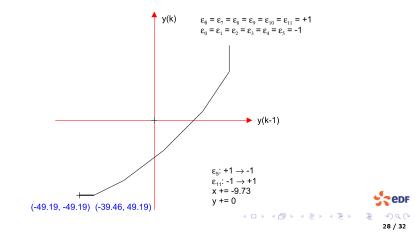


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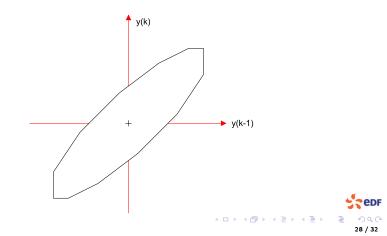
$$\begin{cases} y(k) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) \in \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{cases}$$



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$$\begin{array}{rcl} y(k) & \in & \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k} \cos\left(\frac{\pi}{3} - k\frac{\pi}{6}\right) \varepsilon_{k} \\ y(k-1) & \in & \frac{\|u\|_{\infty}}{1 - \frac{3^{6}}{2^{12}}} \sum_{k=0}^{11} \left(\frac{\sqrt{3}}{2}\right)^{k-1} \cos\left(\frac{\pi}{2} - k\frac{\pi}{6}\right) \varepsilon_{k} \end{array}$$



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Référe	ences l			

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Référe	nces III			

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Context	1 st order filter	filter with real eigenvalues	filter with complex eigenvalues	Conclusion
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Référe	nces IV			

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